



COST Action IC 1407 – Training School

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REVERBERATION CHAMBERS

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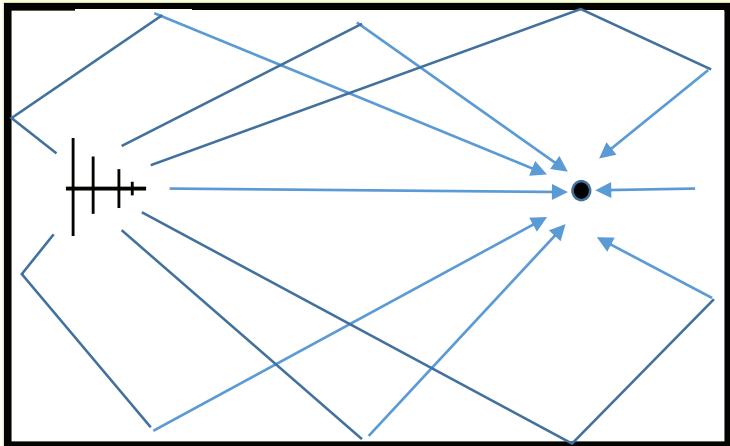
Definition of Reverberation Chamber

A properly operating RC is an electrically large cavity, where the electromagnetic field is statistically uniform, isotropic and randomly polarised within an acceptable and predictable uncertainty and confidence limit

- Uniformity implies all spatial locations within RC (at sufficient distance from metal surfaces) are equivalent
- Isotropic implies that at given location in RC electromagnetic energy is same in any direction
- Random polarization implies that the phase relationships between polarized components are equivalent

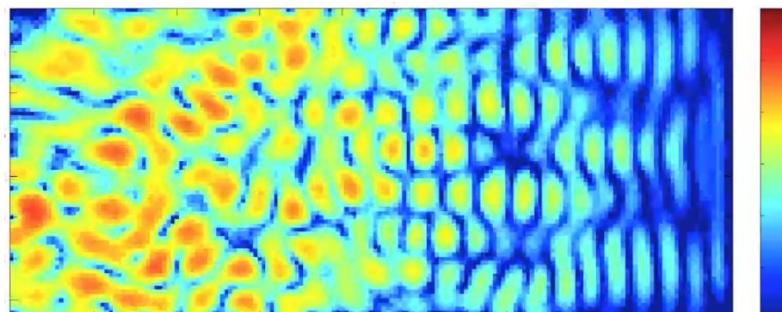
.....electrically large cavity...

Time Domain



$$\vec{e}_T = \sum_{n=1}^{\infty} \vec{e}_n (t - \tau_n) u(t - \tau_n)$$

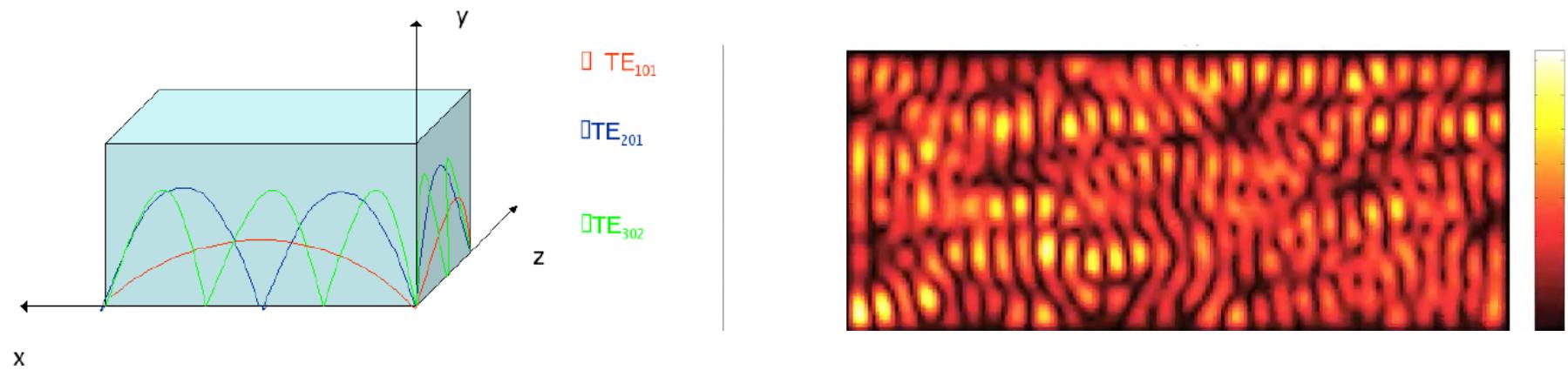
$$u(t - \tau_n) = \begin{cases} 1 & \text{if } t \geq \tau_n \\ 0 & \text{if } t < \tau_n \end{cases}$$



.....electrically large cavity...

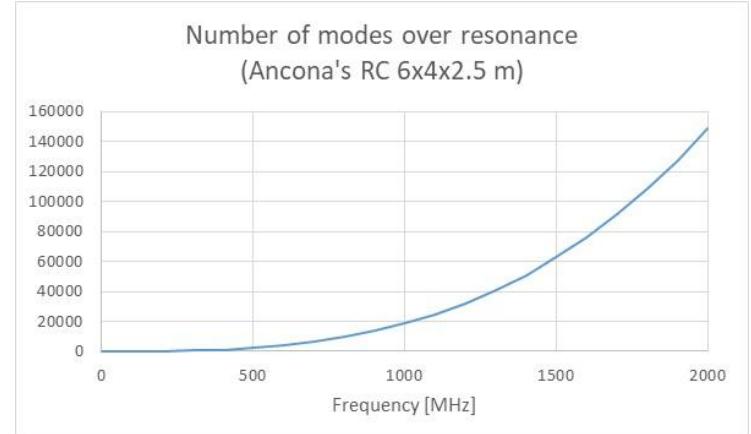
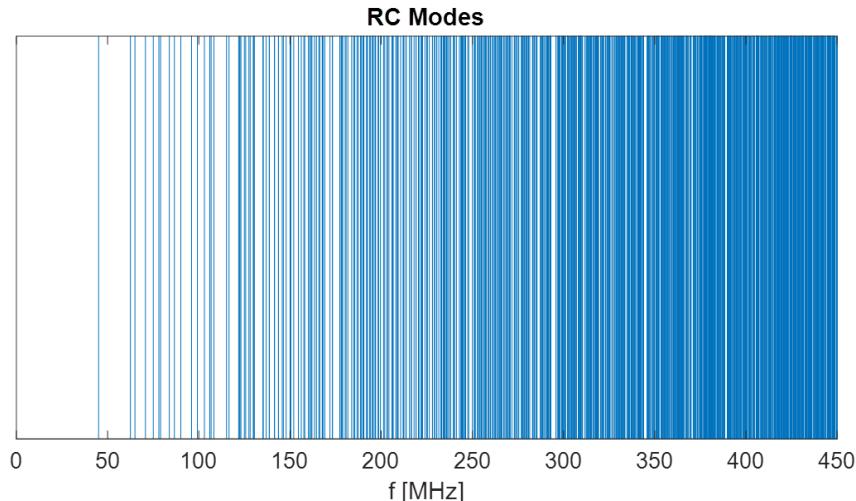
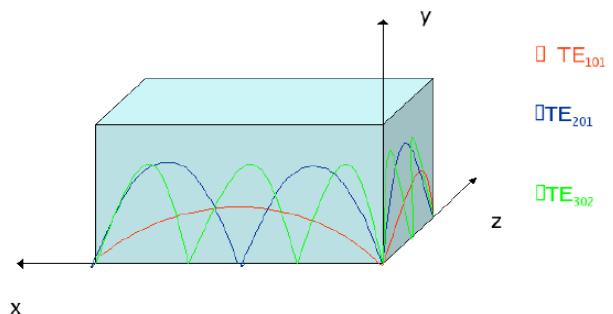
Frequency Domain

$$\vec{E} = -\frac{1}{j\omega\epsilon} \sum_{m=0}^{\infty} \sum_{n=0}^{\infty} \sum_{p=0}^{\infty} \frac{\int_V \vec{J} \cdot \vec{f}_{mnp} dV}{\int_V |\vec{f}_{mnp}|^2 dV} \vec{f}_{mnp} - j\omega\mu \sum_{m=0}^{\infty} \sum_{n=0}^{\infty} \sum_{p=0}^{\infty} \frac{\int_V \vec{J} \cdot \vec{e}_{mnp} dV}{(\vec{k}_{mnp}^2 - \tilde{k}^2) \int_V |\vec{e}_{mnp}|^2 dV} \vec{e}_{mnp}$$



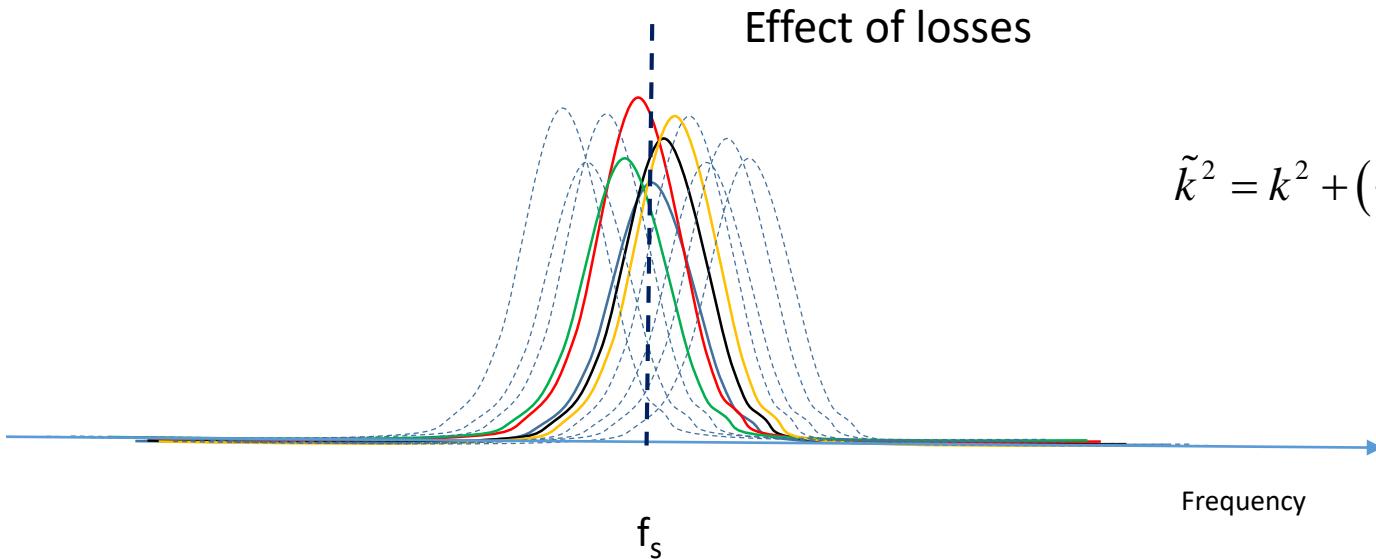
Cavity modes

$$f_{mnp} = \frac{c}{2} \sqrt{\left(\frac{m}{a}\right)^2 + \left(\frac{n}{b}\right)^2 + \left(\frac{p}{d}\right)^2}$$



$$N = \frac{8\pi}{3} V \left(\frac{f}{c} \right)^3 - (a + b + d) \left(\frac{f}{c} \right) + \frac{1}{2}$$

Cavity modes



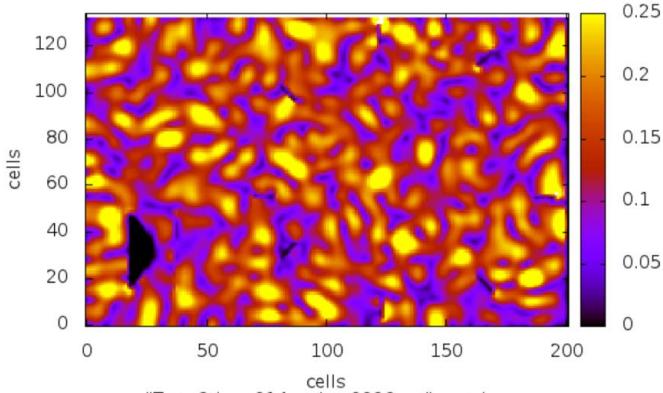
$$\tilde{k}^2 = k^2 + (-1 + j) \frac{k^2 \omega_m}{Q_m \omega}$$

$$\frac{dN}{df} = \frac{8\pi V}{c^3} f^2 \quad BW \approx f/Q$$

$$N_S = \frac{8\pi V f^3}{c^3 Q}$$

Rule of thumb: at least 60 modes over resonance (typically it happens between 3 F_o and 6 F_o)

.....statistically uniform, isotropic and randomly polarised



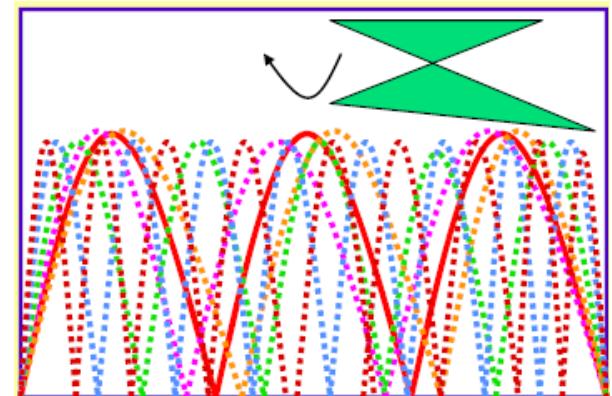
For TE_{mnp} $E_z=0$
 For TM_{mnp} $H_z=0$



For a single chamber state

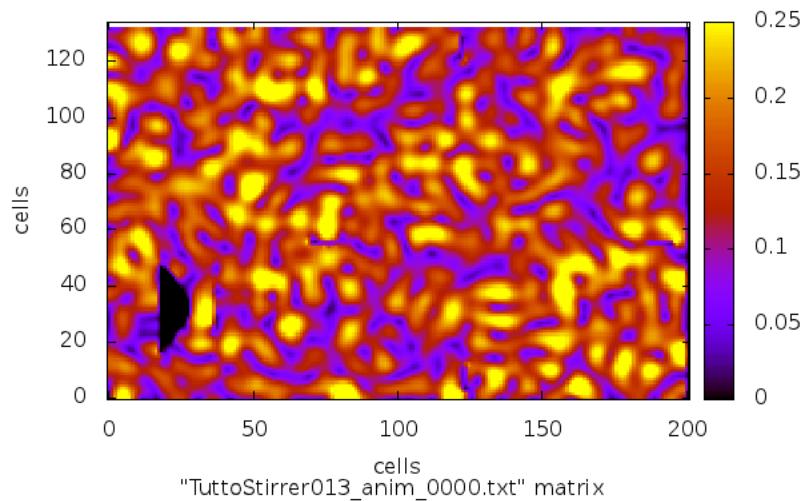
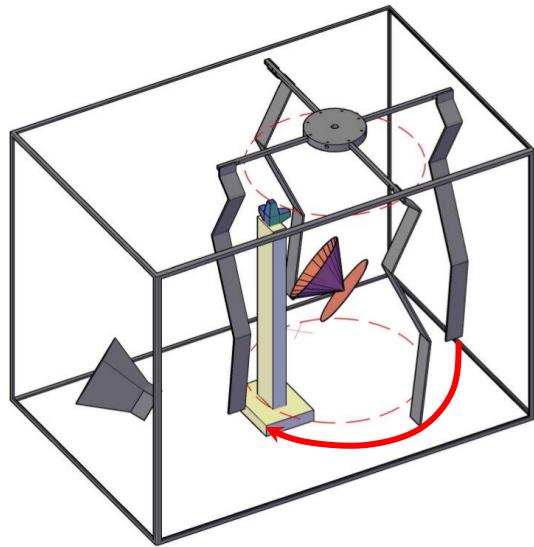
- No uniformity
- No isotropy
- No random polarization

$$\vec{E} = \dots - j\omega\mu \sum_{m=0}^{\infty} \sum_{n=0}^{\infty} \sum_{p=0}^{\infty} \frac{\int_V \vec{J} \cdot \vec{e}_{\text{mnp}} dV}{(k_{\text{mnp}}^2 - \tilde{k}^2) \int_V |\vec{e}_{\text{mnp}}|^2 dV} \vec{e}_{\text{mnp}}$$



We need a stirring action to get a lot of different chamber realizations

Example of mechanical stirring

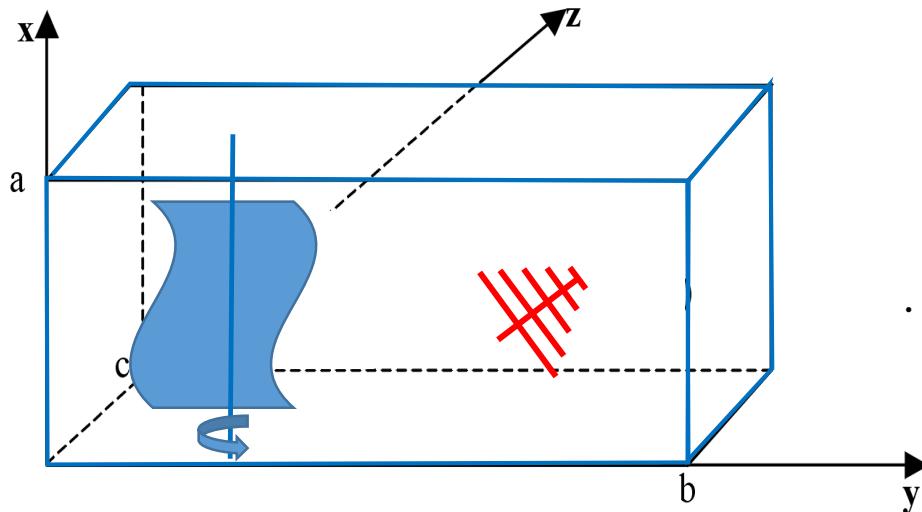


F. Moglie and V. Mariani Primiani, "Numerical Analysis of a New Location for the Working Volume Inside a Reverberation Chamber," in *IEEE Transactions on Electromagnetic Compatibility*, vol. 54, no. 2, pp. 238-245, April 2012.

Sample #	Prec.	Ex	Ey	Ez
1				
2				
3				
4				
5				
.....				
.....				
.....				
N				

Ensemble average , mean value ($\langle X \rangle_N$)
Maximum value
Max/Mean ratio
Statistical distributions: PDF and CDF.

Field stirring actions



Source stir.

$$\cdots j\omega\mu \sum_{m=0}^{\infty} \sum_{n=0}^{\infty} \sum_{p=0}^{\infty} \left(k_{mnp}^2 - \tilde{k}^2 \right) \int_V \left| \vec{e}_{mnp} \right|^2 dV$$

Mechanical

$$\int_V \vec{J} \cdot \vec{e}_{mnp} dV$$

$$\vec{e}_{mnp}$$

MECHANICAL STIRRING, i.e. boundary condition variations

- Rotating paddles (NIST group,)
- Moving walls (Capsalis, ...)
- Vibrating walls (VIRC by Leferink, ...)
-

For a review and history of all techniques see R. Serra, A. Marvin, F. Leferink, V. Mariani Primiani, F. Moglie, M. O. Hatfield, Y. Huang, L. Arnaut, A. Cozza, M. Klinger, "Reverberation chambers a la carte: An overview of the different mode-stirring techniques," in *IEEE Electromagnetic Compatibility Magazine*, vol. 6, no. 1, pp. 63-78, First Quarter 2017.

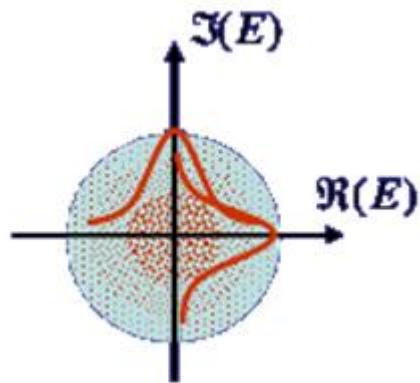
SOURCE STIRRING

- Generator amplitude or phase variation (Hill, ...)
- Frequency variation (frequency stirring) (T. A. Loughry, ...)
- Moving transmitting antenna to change mode coupling (Huang, Carlberg, Kildal, ...)
-

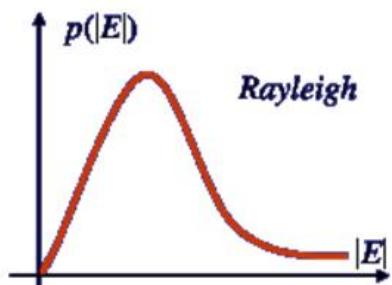
Statistical distributions in an ideal RC

$$E_x = E_{xr} + iE_{xi}, \quad E_y = E_{yr} + iE_{yi}, \quad E_z = E_{zr} + iE_{zi}$$

$$\langle E_{xr} \rangle = \langle E_{xi} \rangle = \langle E_{yr} \rangle = \langle E_{yi} \rangle = \langle E_{zr} \rangle = \langle E_{zi} \rangle = 0$$



$$f(E_{xr}) = \frac{1}{\sqrt{2\pi}\sigma} \exp\left[-\frac{E_{xr}^2}{2\sigma^2}\right]$$



CHI-2DOF (χ_2)

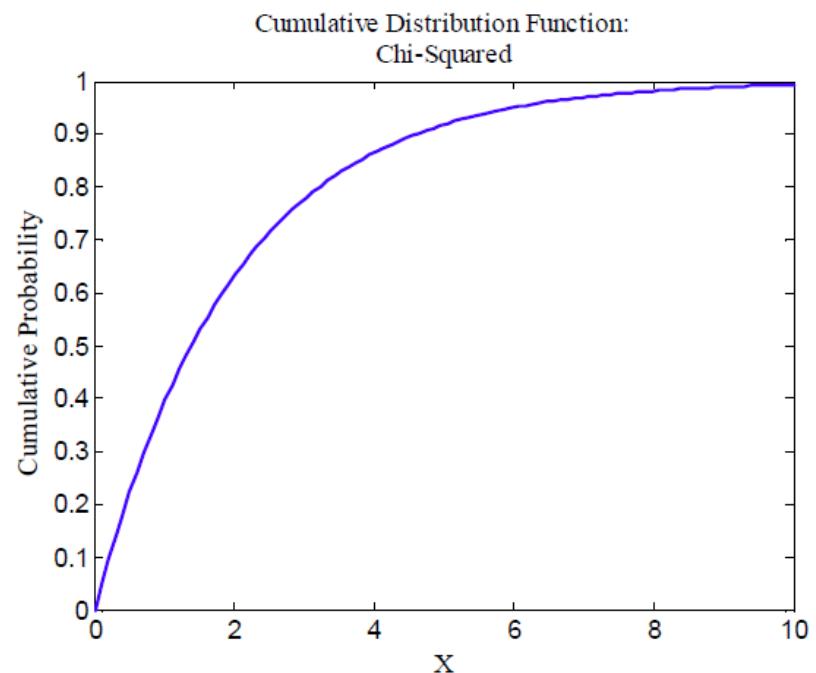
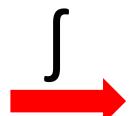
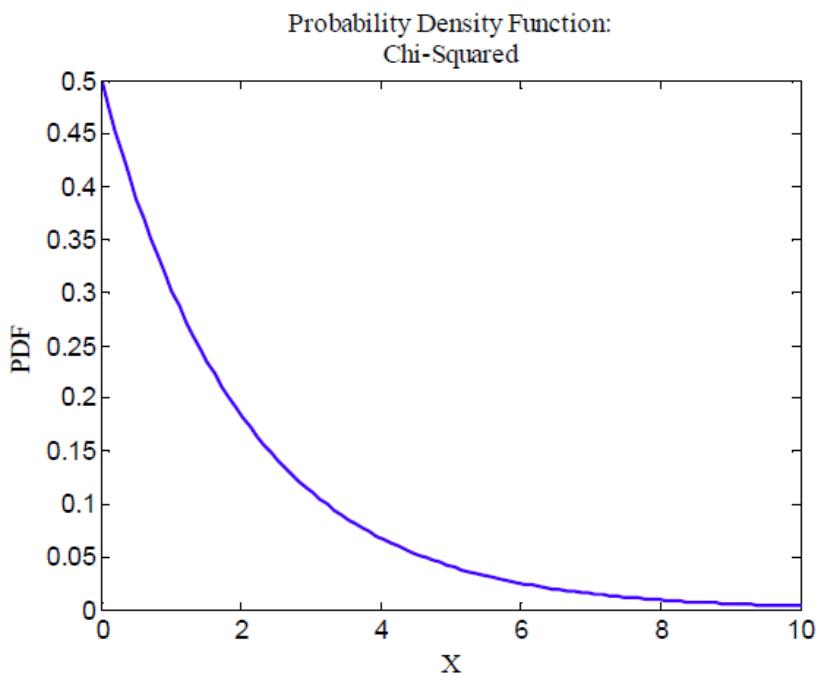
$$f(|E_x|) = \frac{|E_x|}{\sigma^2} \exp\left[-\frac{|E_x|^2}{2\sigma^2}\right]$$

Distribution of the received power P_R or of $|E_i|^2$

CHI²-2DOF

χ^2_2

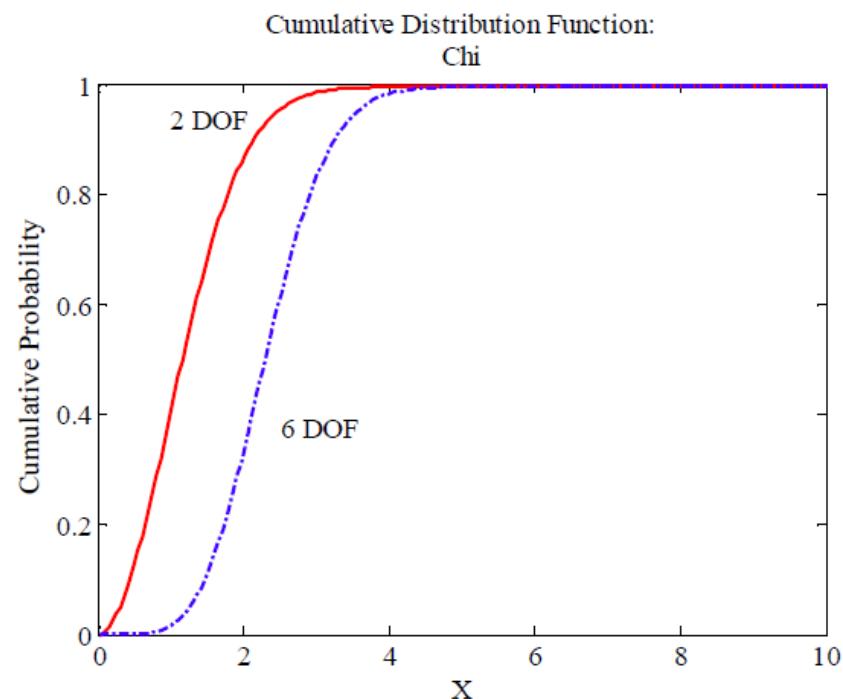
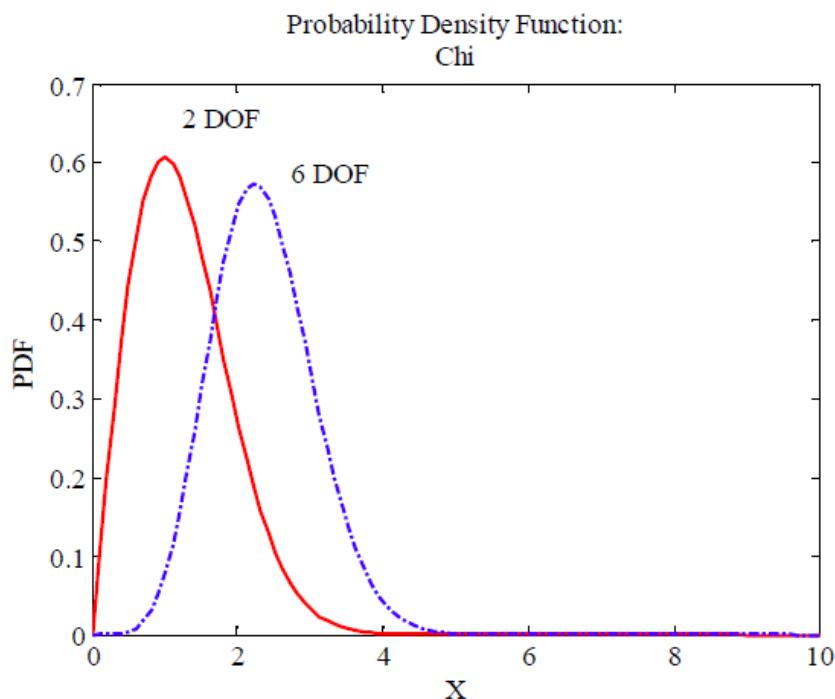
$$f(|E_x|^2) = \frac{1}{2\sigma^2} \exp\left[-\frac{|E_x|^2}{2\sigma^2}\right]$$



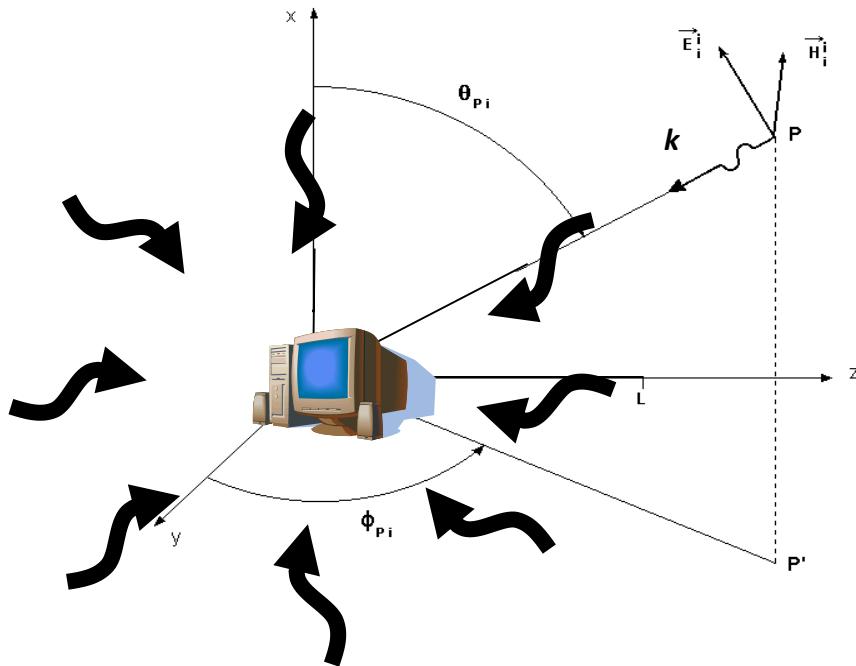
Statistical distribution of the total E-field

CHI-6DOF χ_6

$$f(|\mathbf{E}|) = \frac{|\mathbf{E}|^5}{8\sigma^6} \exp\left[-\frac{|\mathbf{E}|^2}{2\sigma^2}\right]$$



An elegant view of the ideal RC field: the plane wave integral representation (by Hill)



$$E(\mathbf{r}) = \iint_{4\pi} F(\Omega) \exp(i\mathbf{k} \cdot \mathbf{r}) d\Omega.$$

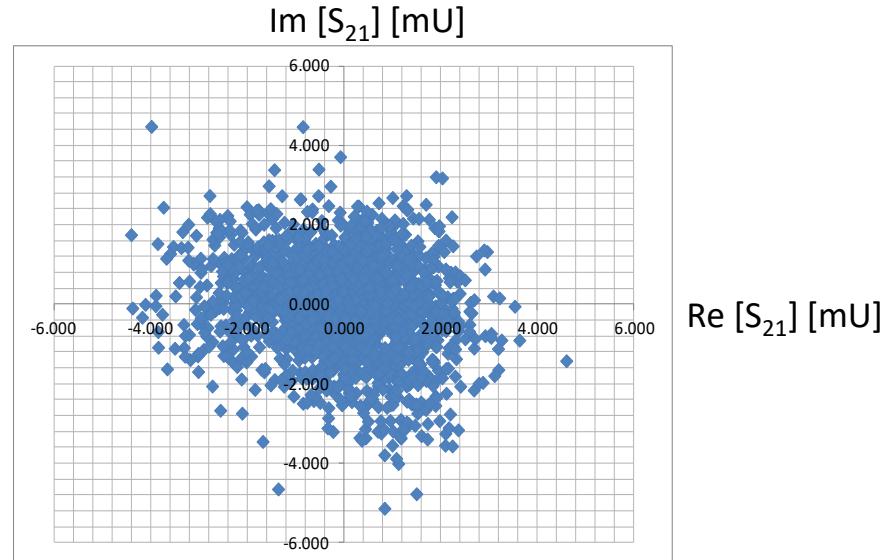
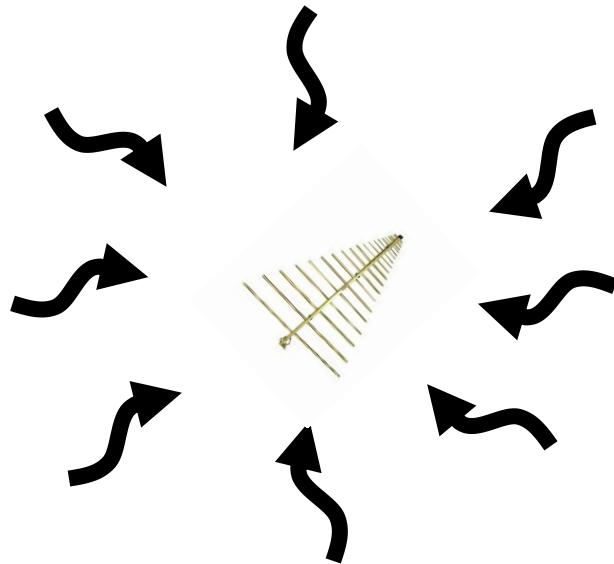
$$\mathbf{k} = -k(\hat{x} \sin \alpha \cos \beta + \hat{y} \sin \alpha \sin \beta + \hat{z} \cos \alpha).$$

$$\langle E(\mathbf{r}) \rangle = \iint_{4\pi} \langle F(\Omega) \rangle \exp(i\mathbf{k} \cdot \mathbf{r}) d\Omega = 0$$

$$|E(\mathbf{r})|^2 = \iint_{4\pi} \iint_{4\pi} F(\Omega_1) \bullet F^*(\Omega_2) \exp[i(\mathbf{k}_1 - \mathbf{k}_2) \bullet \mathbf{r}] d\Omega_1 d\Omega_2 \equiv E_0^2$$

$$S = \frac{E_0^2}{377} \quad \text{Scalar Power Density}$$

Antenna behaviour in ideal RC



$$\langle D \rangle = \langle g_D(\theta, \varphi) \rangle = 1$$

$$\langle A_{\text{eff}} \rangle = \frac{1}{2} \frac{\lambda^2}{4\pi}$$

Free space $\frac{A_{\text{eff}}}{D}$

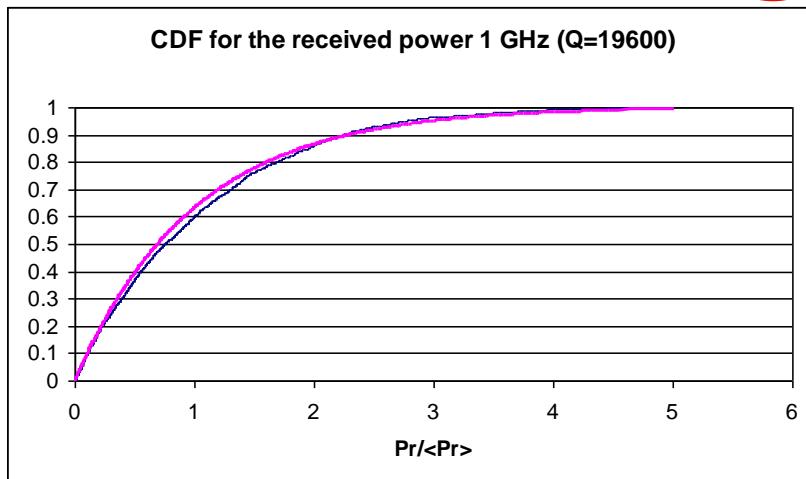
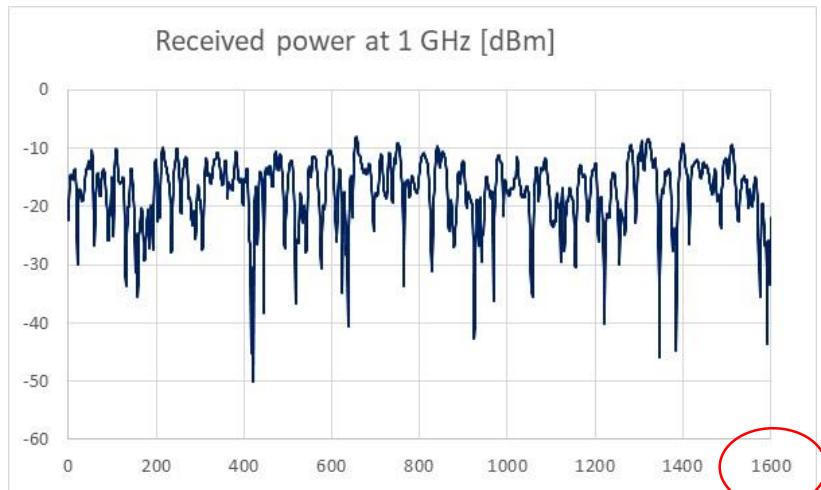
$$\langle P_R \rangle = S \cdot \langle A_{\text{eff}} \rangle \cdot m \cdot \eta_{RX} = S \cdot \frac{\lambda^2}{8\pi} \cdot m \cdot \eta_{RX}$$

Polarization mismatching

$$m = \langle (1 - |S_{11}|^2) \rangle \approx (1 - |\langle S_{11} \rangle|^2)$$

η_{RX} Antenna efficiency (effect of antenna losses)

Example of received power statistics

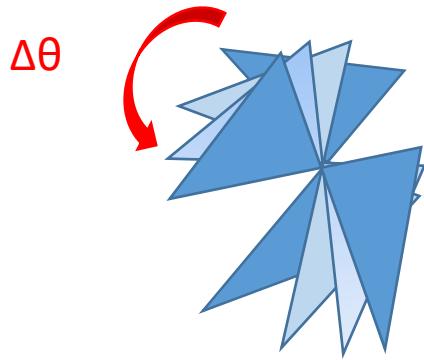


Number of independent samples $N_{\text{ind}} = 200$

$$CDF \left\{ \chi^2_2(s) \right\} = 1 - \exp \{-s\}$$

s = mean normalised
received power

Determination of uncorrelated stirrer positions



$$r = \frac{\frac{1}{n-1} \sum_i^n (x_i - u_x)(y_i - u_y)}{\sqrt{\left(\frac{\sum_i^n (x_i - u_x)^2}{n-1} \right) \left(\frac{\sum_i^n (y_i - u_y)^2}{n-1} \right)}}$$

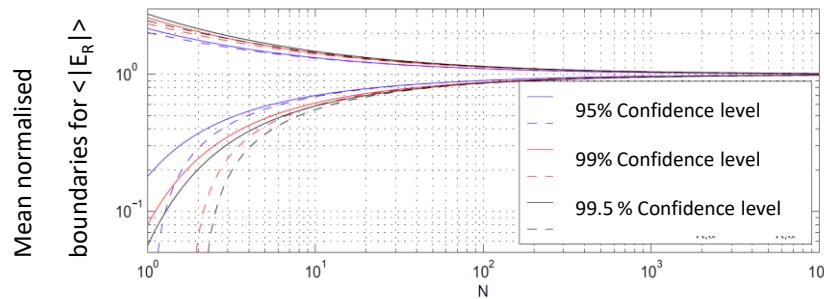
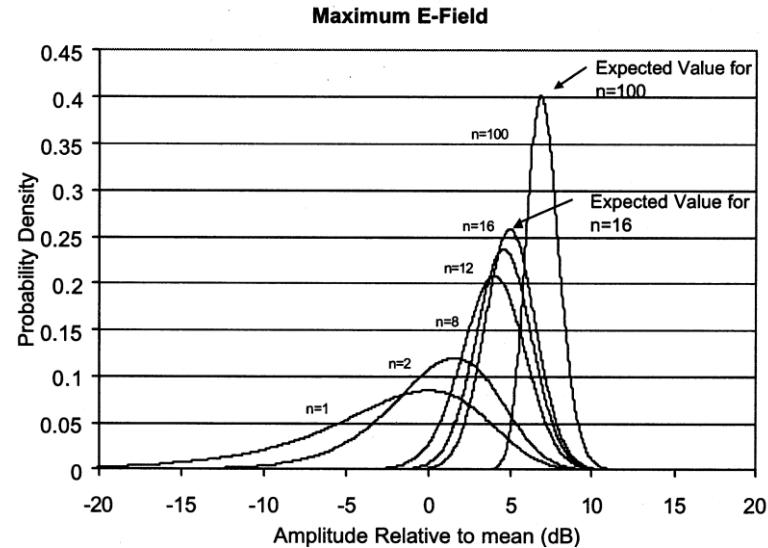
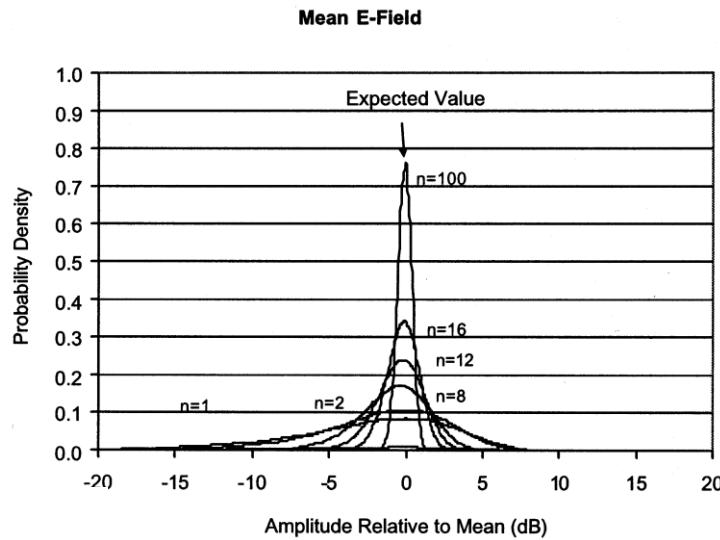
$$r \approx 0.37 \cdot \left(1 - \frac{7.22}{n^{0.64}} \right)$$

Circular autocorrelation

										M = 1601/8	200	
	shift 1	shift 2	shift 3	shift 4	shift 5	shift 6	shift 7	shift 8	shift 9	shift 10		
0.968247	0.885106	0.774307	0.656979	0.546131	0.447477	0.362268	0.289546	0.22754	0.1746			
0.07587	0.079966	0.048021	0.024816	0.021428	0.036721	0.049407	0.051975	0.043173	0.024282	0.006603		
0.11349	0.07587	0.079966	0.048021	0.024816	0.021428	0.036721	0.049407	0.051975	0.043173	0.024282		
0.144328	0.11349	0.07587	0.079966	0.048021	0.024816	0.021428	0.036721	0.049407	0.051975	0.043173		
0.169102	0.144328	0.11349	0.07587	0.079966	0.048021	0.024816	0.021428	0.036721	0.049407	0.051975		
0.18441	0.169102	0.144328	0.11349	0.07587	0.079966	0.048021	0.024816	0.021428	0.036721	0.049407		
0.18675	0.18441	0.169102	0.144328	0.11349	0.07587	0.079966	0.048021	0.024816	0.021428	0.036721		
"	"	"	"	"	"	"	"	"	"	"		
"	"	"	"	"	"	"	"	"	"	"		
"	"	"	"	"	"	"	"	"	"	"		
"	"	"	"	"	"	"	"	"	"	"		
"	"	"	"	"	"	"	"	"	"	"		
0.049407	0.051975	0.043173	0.024282	0.006603	0.029461	0.046321	0.049953	0.047316	0.046356	0.054704		
0.036721	0.049407	0.051975	0.043173	0.024282	0.006603	0.029461	0.046321	0.049953	0.047316	0.046356		
0.021428	0.036721	0.049407	0.051975	0.043173	0.024282	0.006603	0.029461	0.046321	0.049953	0.047316		
0.024816	0.021428	0.036721	0.049407	0.051975	0.043173	0.024282	0.006603	0.029461	0.046321	0.049953		
0.048021	0.024816	0.021428	0.036721	0.049407	0.051975	0.043173	0.024282	0.006603	0.029461	0.046321		
0.079966	0.048021	0.024816	0.021428	0.036721	0.049407	0.051975	0.043173	0.024282	0.006603	0.029461		

Stirrer independent (uncorrelated) positions

.....within an acceptable and predictable uncertainty and confidence limit.



$$PDF(r) = M \frac{\pi}{2} r \left[1 - e^{\left(-\frac{\pi}{4} r^2 \right)} \right]^{M-1} e^{\left(-\frac{\pi}{4} r^2 \right)} \quad r = \frac{E_{\max}}{\langle E \rangle}$$

See Appendix K of IEC 61000-4-21 and related references for uncertainty as function of N and confidence level

Chamber quality factor

$$Q = \frac{\omega U}{P_D} = \frac{1}{\frac{1}{Q_w} + \frac{1}{Q_{Ant}} + \frac{1}{Q_{Abs}} + \frac{1}{Q_{Other}}}$$

$$Q_w \approx \frac{3V}{2\mu_r \delta A} \propto \sqrt{f}$$

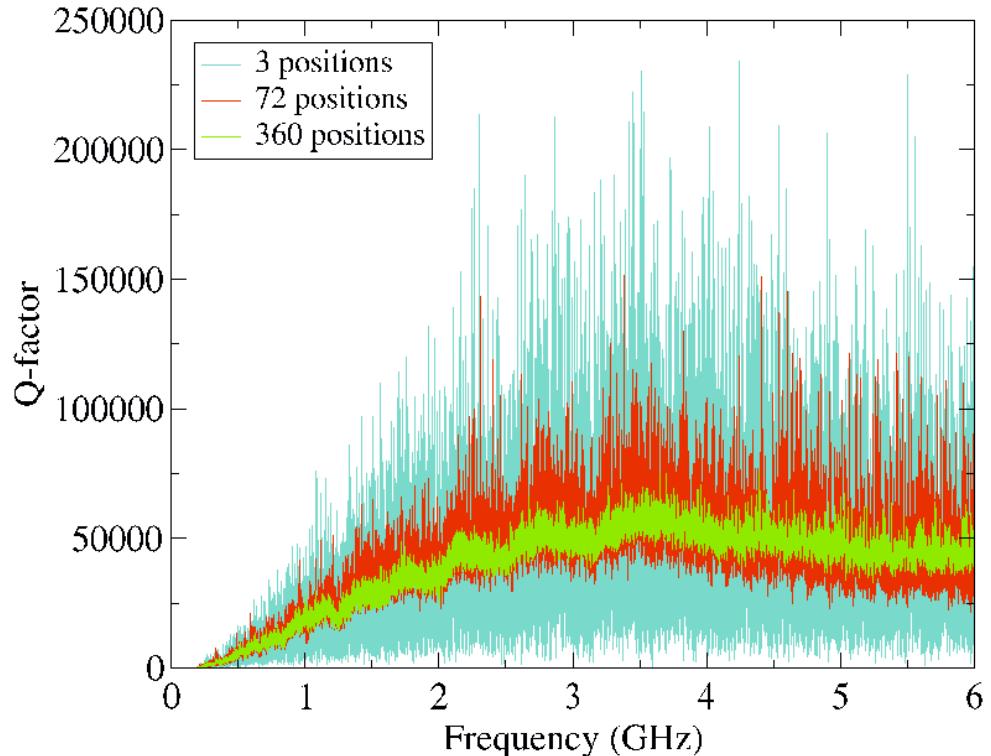
$$\delta = \sqrt{\pi f \sigma \mu}$$

$$Q_{Ant} = \frac{16\pi^2 V}{m\lambda^3} \propto f^3$$

$$Q_{Abs} = \frac{2\pi V}{\lambda \langle \sigma_a \rangle_\Omega} \propto f$$

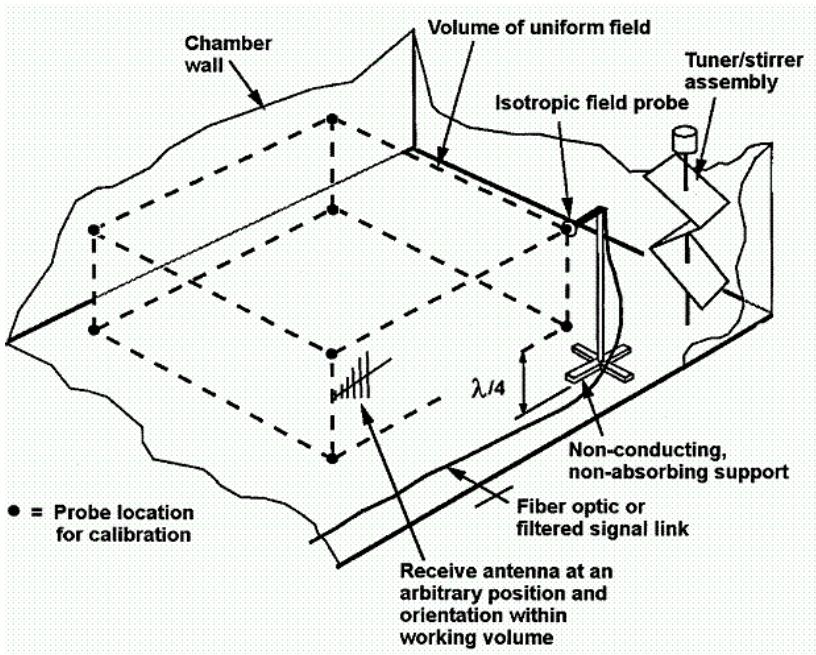
Measured Q

$$Q = \frac{16\pi^2 V}{\eta_{Tx}\eta_{Rx}\lambda^3} \left\langle \frac{P_{AveRec}}{P_{Input}} \right\rangle_n$$



Another chamber performance indicator: the field uniformity

$$\vec{E}_{x,y,z} = \frac{E_{MAX\ x,y,z}}{\sqrt{P_{INPUT}}} ,$$



$$\langle \vec{E}_x \rangle_8 = (\sum \vec{E}_x) / 8$$

$$\langle \vec{E}_y \rangle_8 = (\sum \vec{E}_y) / 8$$

$$\langle \vec{E}_z \rangle_8 = (\sum \vec{E}_z) / 8$$

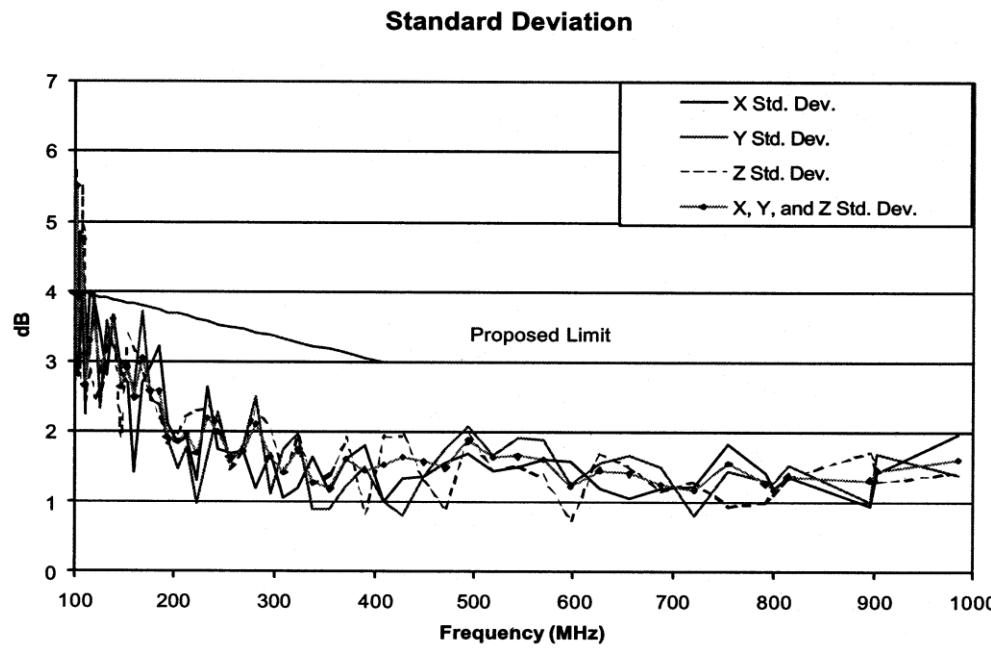
$$\langle \vec{E} \rangle_{24} = (\sum \vec{E}_{x,y,z}) / 24,$$

Field uniformity in the working volume

$$\sigma_x = \sqrt{\frac{\sum (E_{ix} - \langle E_x \rangle_8)^2}{8-1}}$$

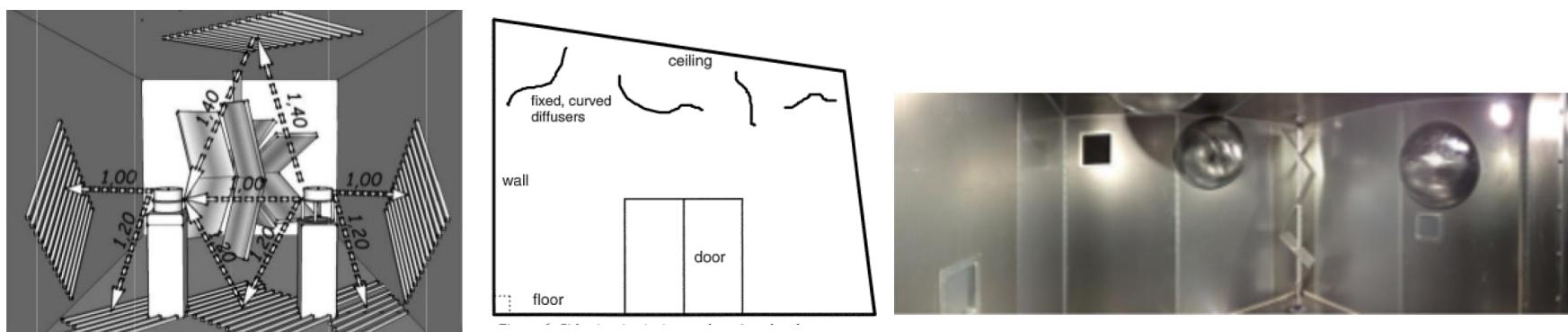
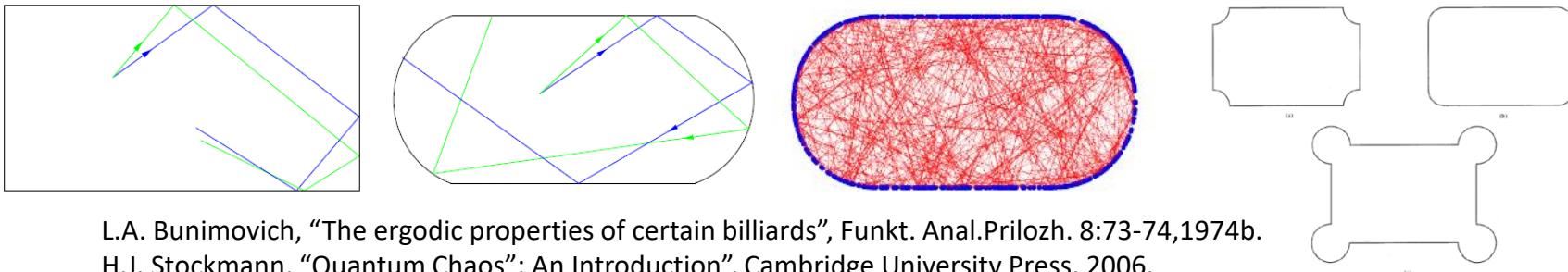
$$\sigma_{24} = \sqrt{\frac{\sum_{m=1}^8 \sum_{n=1}^3 (E_{m,n} - \langle E \rangle_{24})^2}{24-1}}$$

$$\sigma(db) = 20 \cdot \log \left(\frac{\sigma + \langle E_{xyz} \rangle}{\langle E_{xyz} \rangle} \right)$$



IEC 61000-4-21

Improvements of Chamber performance



F. Leferink, "High Field Strength in a Large Volume: The Intrinsic Reverberation Chamber", IEEE EMC 1998, Denver, CO, USA

L. Arnaut, "Operation of electromagnetic reverberation chambers with wave diffractors at relatively low frequencies", IEEE Trans. on EMC 2001, pp. 637-653, Vol. 43, No. 4, Nov.

A. C. Marvin, E. Karadimou, "The Use of Wave Diffusers to Reduce the Contribution of Specular Wall Reflections to the Unstirred Energy in a Reverberation Chamber", IEEE EMC 2013, Denver, CO, USA

K. Selemani, J.-B. Gros, E. Richalot, O. Legrand, O. Picon and F. Mortessagne, "Comparison of reverberation chamber shapes inspired from chaotic cavities", IEEE Trans. on EMC 2015, pp. 3-11, Vol. 57, No. 1, Feb

Typical RC applications

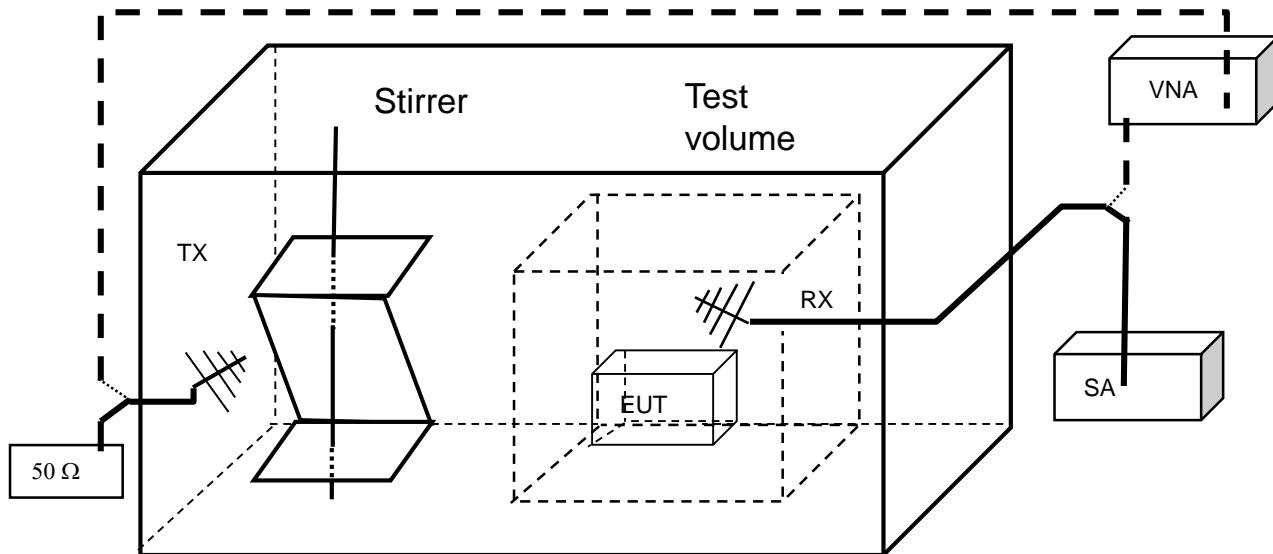
IEC 61000-4-21

Electromagnetic compatibility (EMC) - Part
4-21: Testing and measurement techniques
– Reverberation chamber test methods

Others

- Emission measurements
- Immunity tests
- Shielding effectiveness of
 - Enclosures
 - Cables
 - Gaskets
 - Materials
- Antenna efficiency
- Material characterizations in terms of:
 - Absorbing cross section
 - Scattering cross section
- Testing of wireless devices and systems
 - Multipath propagation
 - Transmission quality
-

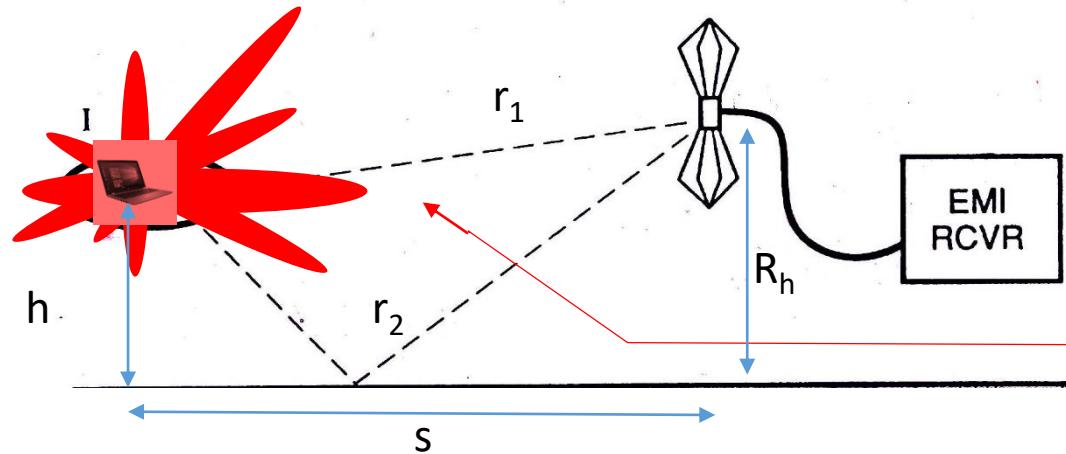
Emission tests



$$P_{RAD} = \frac{\langle P_r \rangle 16\pi^2 V}{\eta_{TX} \eta_{RX} \lambda^3 Q}$$

Follow the procedure in Annex B
of IEC standard to account for Q
variations with the EUT

Correlation with traditional methods



$$E_{\max} = g_{\max} \sqrt{\frac{DP_{RAD} 377}{4\pi R^2}}$$

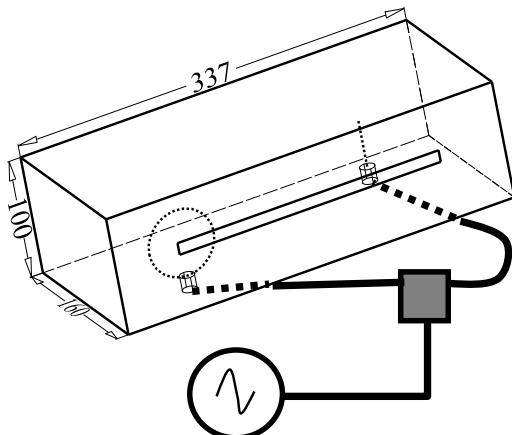
$$g_{\max} = \begin{cases} \left| \frac{r}{r_1} e^{-jkr_1} - \frac{r}{r_2} e^{-jkr_2} \right|_{\max} & \text{for horizontal polarization} \\ \left| \frac{s^2}{r_1^2} \frac{r}{r_1} e^{-jkr_1} + \frac{s^2}{r_2^2} \frac{r}{r_2} e^{-jkr_2} \right|_{\max} & \text{for vertical polarization} \end{cases}$$

D_{MAX} = ?

$$r = \sqrt{s^2 + R_h^2}$$

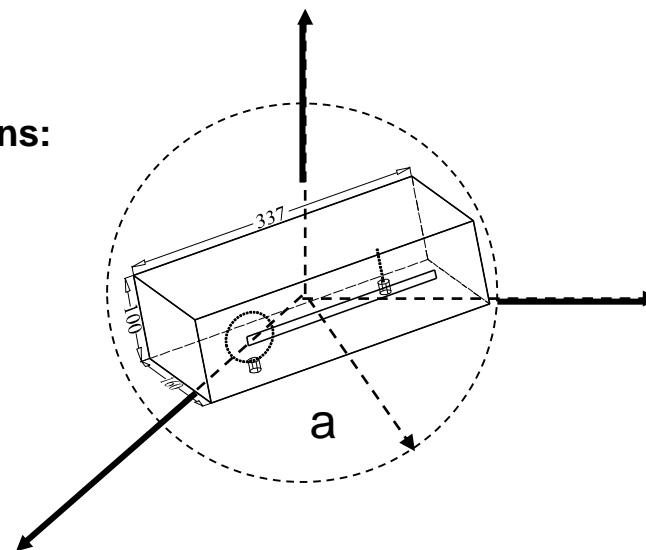
Typically for $s = 10$ m and $1 \leq R_h \leq 4$ m, $g_{\max} \approx 2$

Example



Aperture dimensions:

228 x 2.5 mm

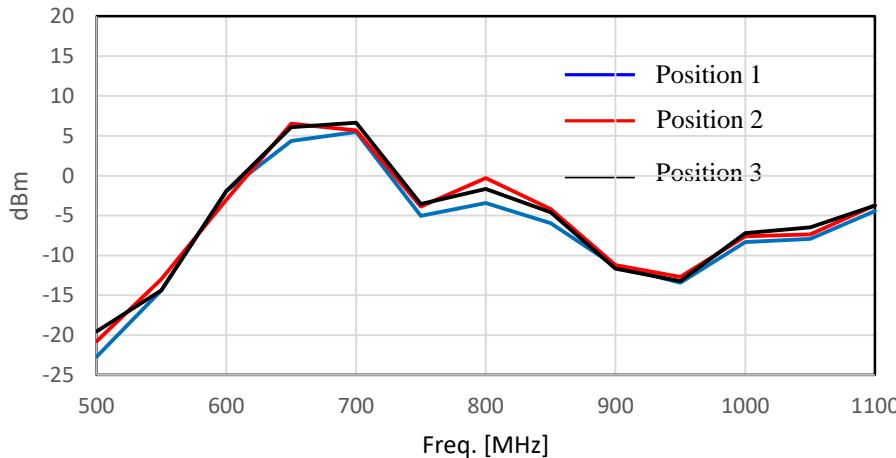


$$D = \begin{cases} 1,55 & \text{per } ka \leq 1 \\ 0,5 \left(0,577 + \ln(4(ka)^2 + 8ka) + \frac{1}{8(ka)^2 + 16ka} \right) & \text{per } ka > 1 \end{cases}$$

MHz	500	550	600	650	700	750	800	850	900	950	1000	1050	1100
D	2,89	2,97	3,04	3,11	3,17	3,23	3,28	3,33	3,38	3,43	3,48	3,52	3,56

Results

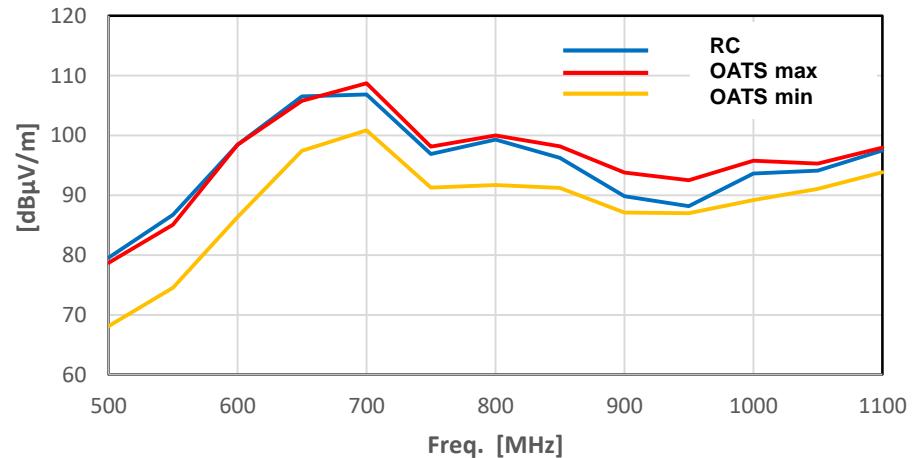
Total radiated power



Independence on EUT positioning
(within the acceptable uniformity)

Good correlation with OATS

Electric field: RC vs. OATS



Immunity tests

$$E_{\text{Eut}} = \left\langle \frac{8\pi}{\lambda} \sqrt{5 \frac{P_{\text{MaxRec}}}{\eta_{\text{rx}}}} \right\rangle_n$$

$$\langle |E_x|_{\text{max}} \rangle \approx \sqrt{\frac{\lambda \eta_v Q}{6\pi V} \left[0,577 \cdot 2 + \ln(N+1) - \frac{1}{2(N+1)} \right]} \langle P_{\text{Tx}} \rangle$$

$$P_{\text{input}} = \left[\frac{E_{\text{test}}}{\langle E \rangle_{24 \text{ or } 9} \sqrt{CLF(f)}} \right]^2$$

Monitor and record:

- $P_{\text{rec. Max}}$ to estimate the maximum field
- $P_{\text{rec.Avg}}$ to control the loading (variations greater than 3 dB must be solved)

During calibration: $P_{\text{in}} = 10 \text{ mW}$, $Q = 5000$, $\langle E_i \rangle = 3.5 \text{ V/m}$

For testing at 150 V/m: $P_{\text{in}} = 18 \text{ W}$ assuming $CLF = 1$

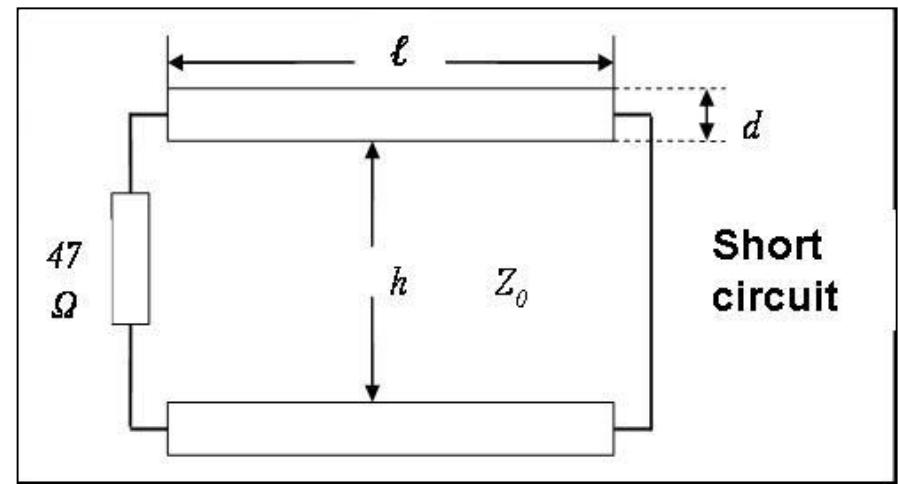
In anechoic chamber $P_{\text{in}} \approx 700 \text{ W}$ (assuming $G= 10 \text{ dB}$ and $r=3 \text{ m}$)

The maximum E field is of interest
for the EUT malfunctioning

$$CLF \leq 1$$

Accounts for quality factor variations w.r.t. calibration

Example: coupling to a transmission line



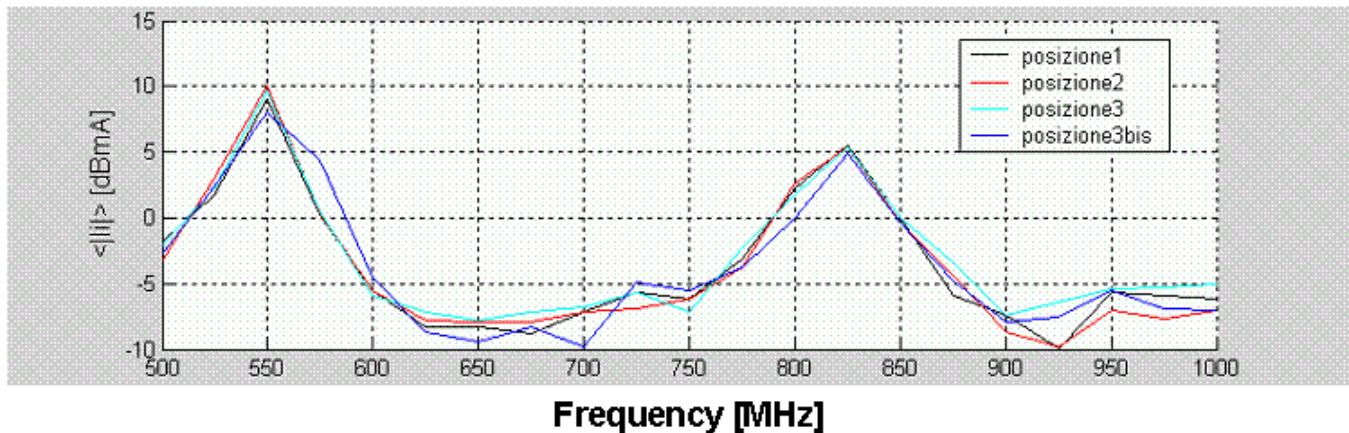
$$\ell = 50 \text{ cm}$$

$$h = 2.5 \text{ cm}$$

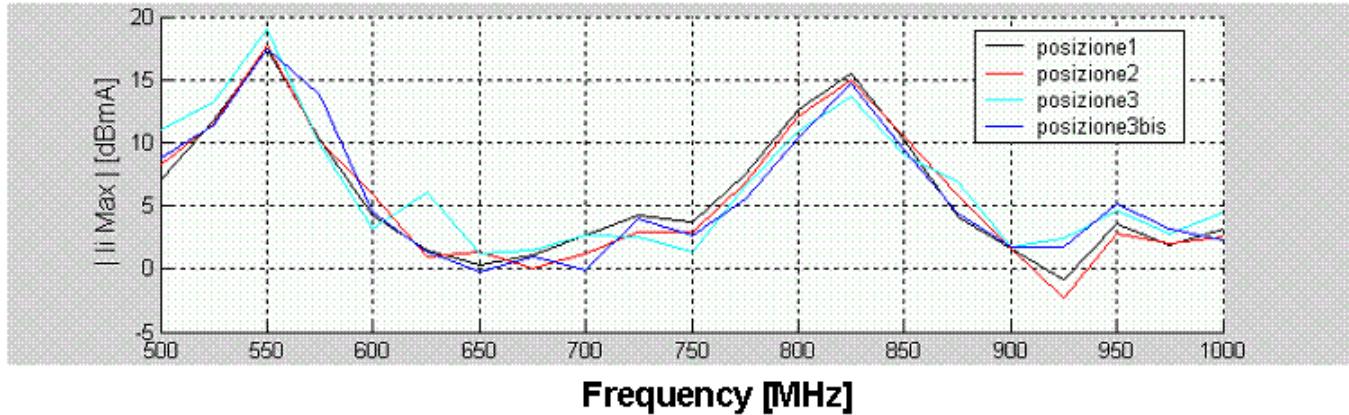
$$Z_0 = 470 \Omega$$

Results: effect of positioning and orientation

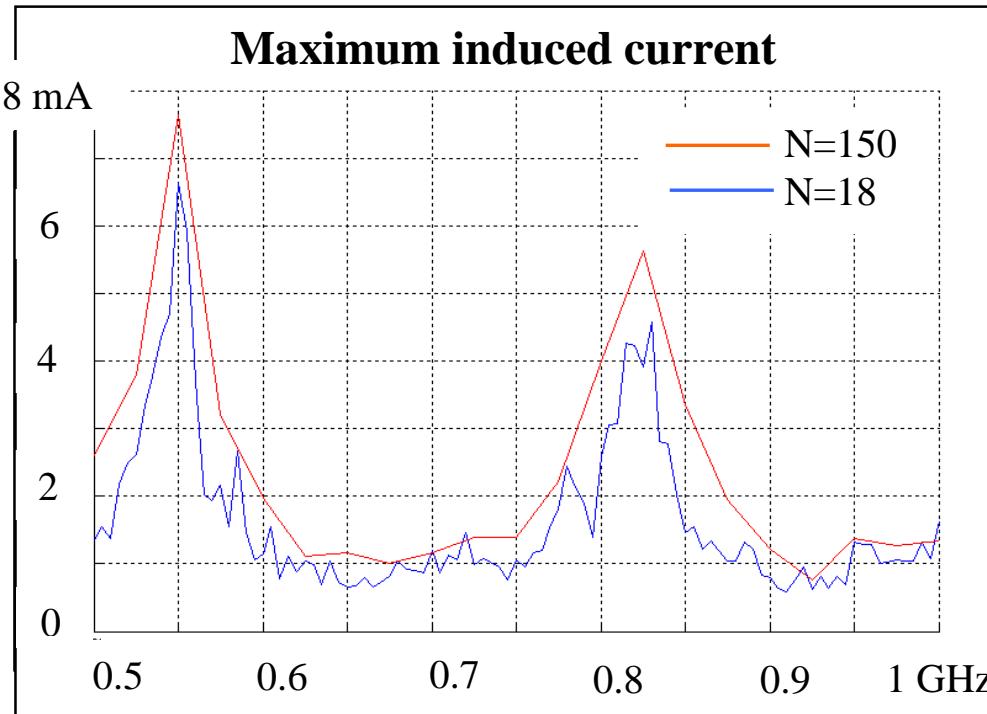
$$\langle |I| \rangle_M$$



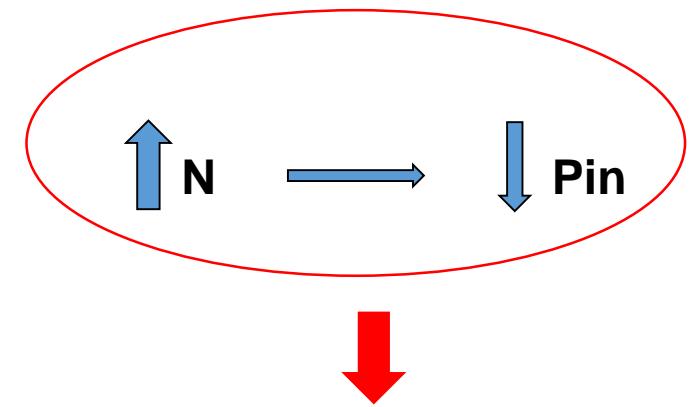
$$|I| \uparrow_M$$



Maximum induced current



$$\text{Max/mean} \sim \ln(N) + 1/2N + 0,577$$



Increasing testing time

Pay attention when you operate in “stirring” w.r.t “tuned” mode

- All independent stirrer positions are employed
- Total stirrer independent positions depend on Q factor

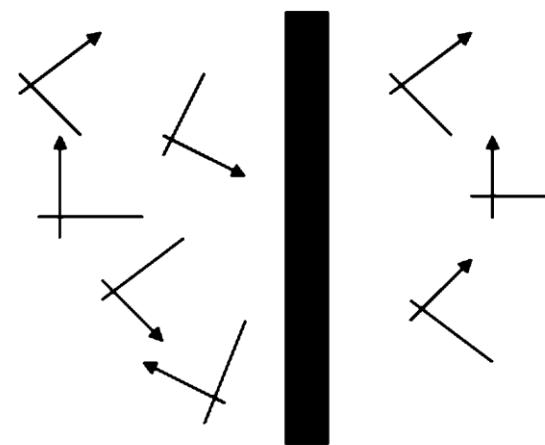
Shielding effectiveness

Conventional methods often uses normal incidence plane wave (e.g. coaxial TEM fixture)



$$SE = -10 \log_{10} \left(\frac{P_t}{P_i} \right)$$

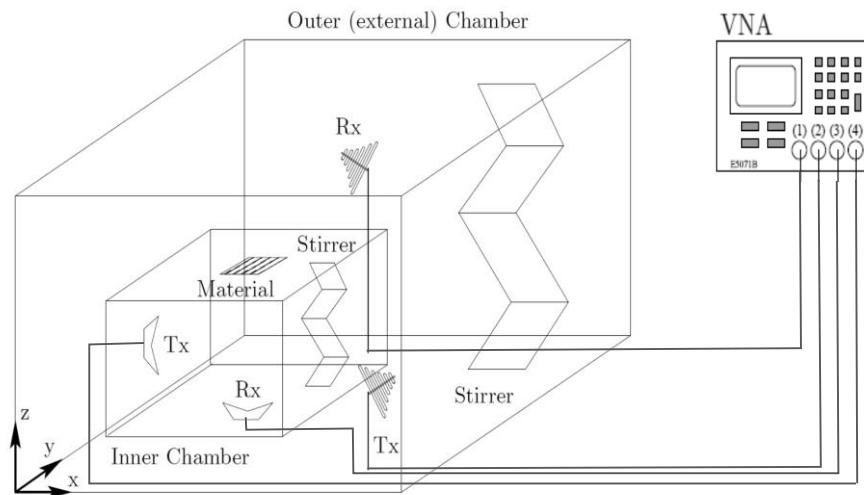
In real complex environments material are exposed to fields that impinge on the material with various polarizations and angle of incidence.



RC field excitation might be more representative of real life situations

Nested reverberation chamber for shielding effectiveness measurements

- The outer chamber provides a **statistically random excitation** of the sample under test
- The inner chamber provides a statistically uniform and isotropic field level due to the energy crossing the sample under test



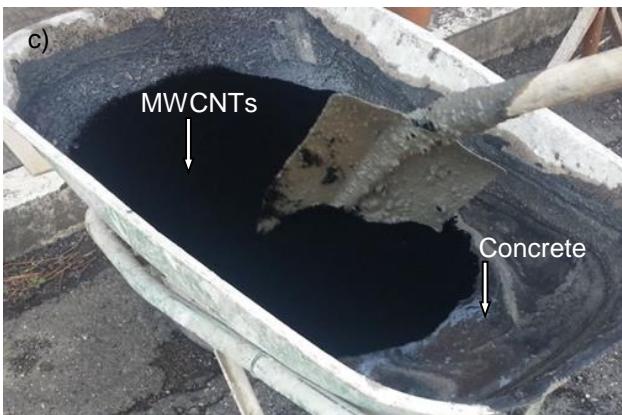
$$Q_{in,ns} = \frac{16\pi^2 V}{\lambda^3} \frac{P_{rQ,in,ns}}{P_{tx,in,ns}}$$

Sample insertion
effect

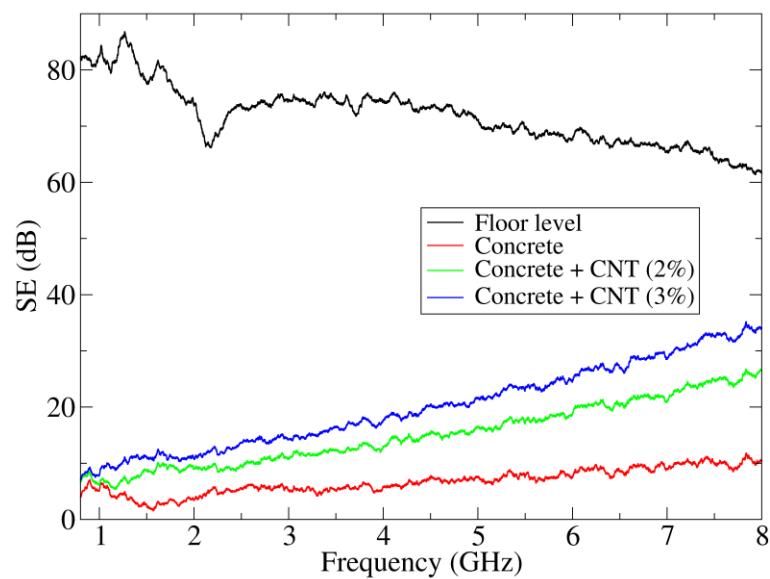
$$Q_{in,s} = \frac{16\pi^2 V}{\lambda^3} \frac{P_{rQ,in,s}}{P_{tx,in,s}}$$

$$SE = -10\log_{10}\left(\frac{P_{r,in,s}}{P_{r,in,ns}} \frac{P_{r,o,ns}}{P_{r,o,s}} \frac{P_{rQ,in,ns}}{P_{rQ,in,s}} \frac{P_{tx,in,s}}{P_{tx,in,ns}}\right)$$

SE of cementitious composites



Pristine MWCNTs
(Nanocyl™ NC 7000,
av. diameter \sim 9.5 nm,
av. length \sim 1.5 μ m,
carbon purity \sim 90%,
surface area 250 \div 300
 m^2/g)

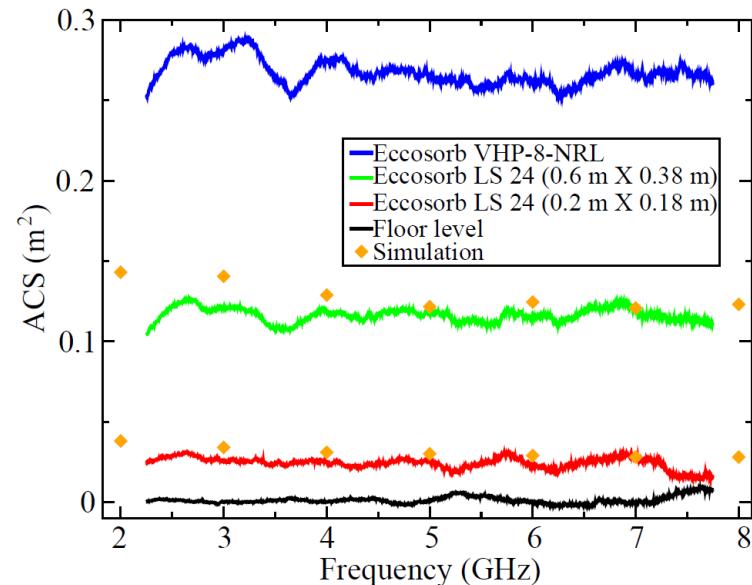
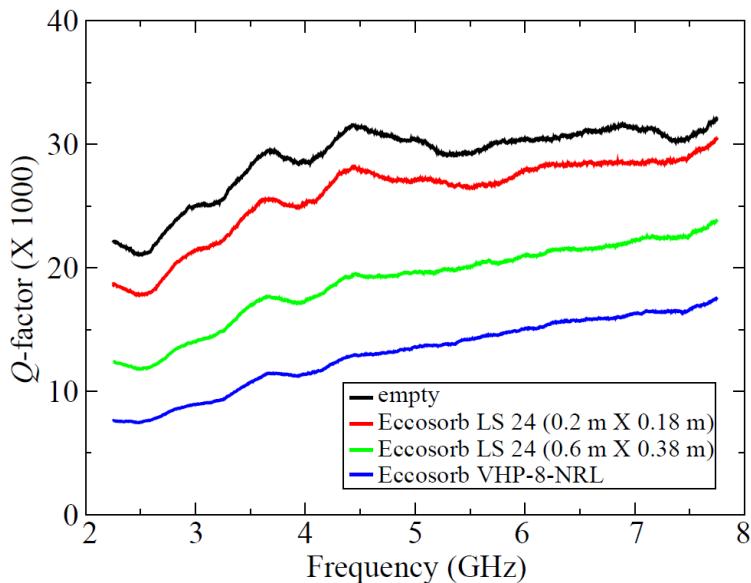


Averaged absorbing cross section (ACS)

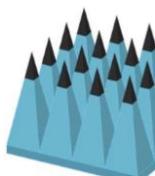
$$ACS = \langle \sigma_{abs} \rangle = \frac{\langle P_{abs} \rangle}{S}$$

$$Q_s = \frac{2\pi V}{\lambda \langle \sigma_a \rangle_\Omega}$$

$$Q_s^{-1} = Q_l^{-1} - Q_u^{-1}$$



G. Gradoni, D. Micheli, F. Moglie, and V. Mariani Primiani,
"Absorbing cross section in reverberation chamber:
experimental and numerical results," *Progress In
Electromagnetics Research B*, Vol. 45, 187-202, 2012.



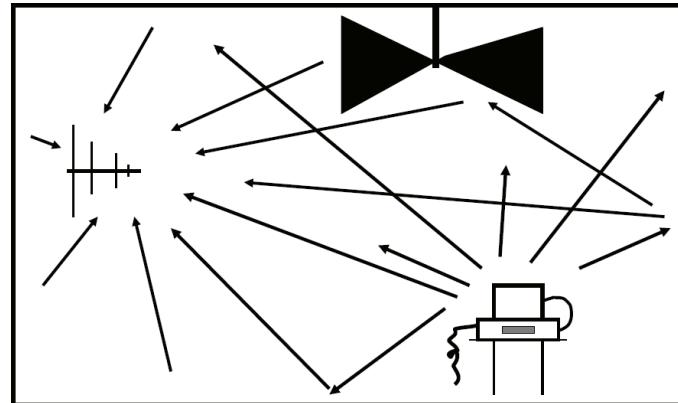
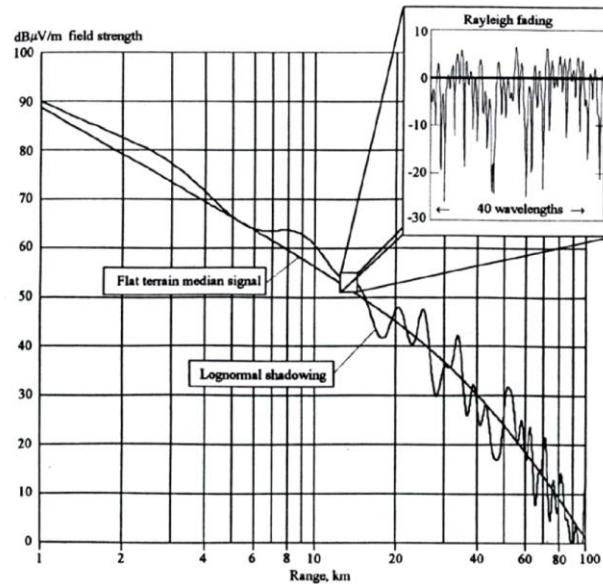
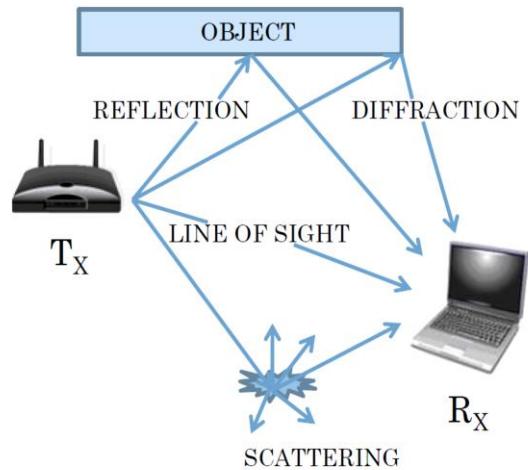
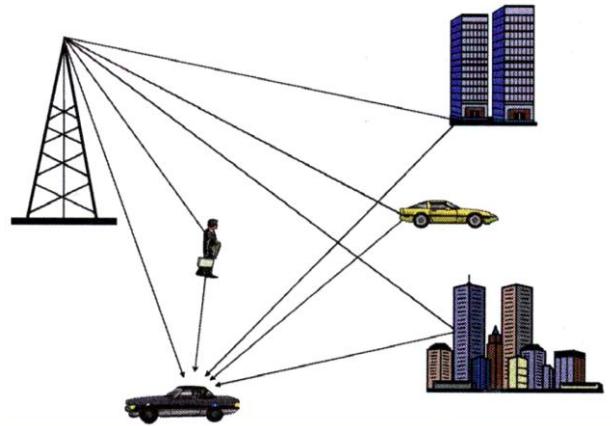
Eccosorb VHP-8-NRL



Eccosorb LS-24L

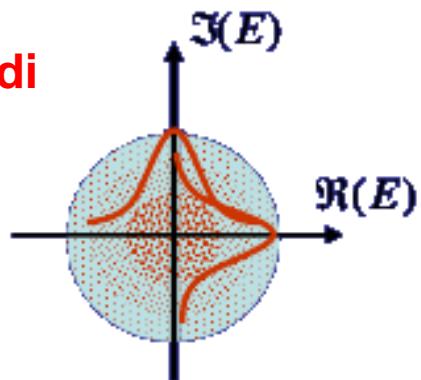
High sensitivity to the insertion of small samples

Testing of wireless devices and systems

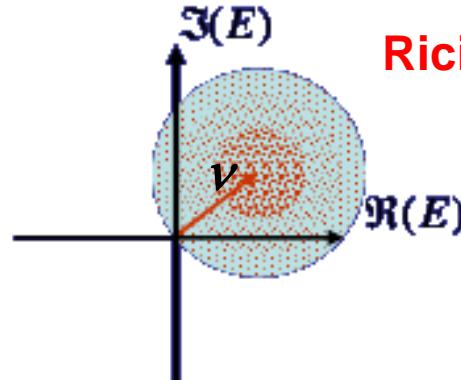


We have to move to a non-ideal RC

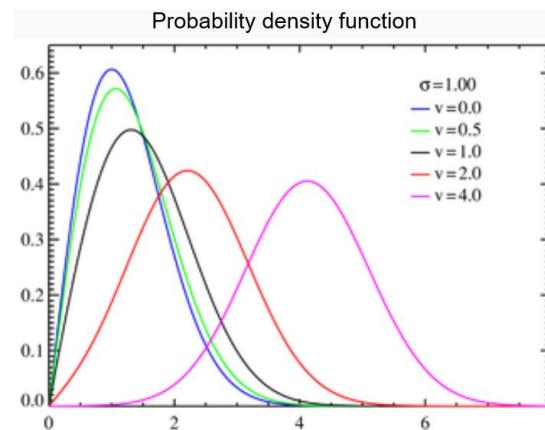
Distribuzione di Rayleigh



Rician distribution



$$f(E_x) = \frac{E_x}{\sigma_x^2} \cdot e^{-\frac{E_x^2}{2\sigma_x^2}}$$



$$K = \frac{\nu^2}{2\sigma^2}$$

$$f(x | \nu, \sigma) = \frac{x}{\sigma^2} \exp\left(\frac{-(x^2 + \nu^2)}{2\sigma^2}\right) I_0\left(\frac{x\nu}{\sigma^2}\right)$$

Testing of wireless devices and systems



ITU Report M.2135-1, Dec. 2009

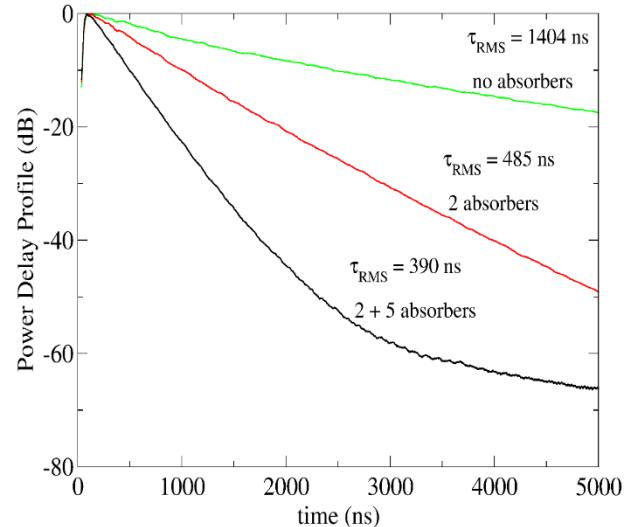
Scenario		Delay spread (ns)
Indoor Hotspot	LOS	20
	NLOS	39
Urban Micro	LOS	65
	NLOS	129
	O-to-I	49
Suburban Macro	LOS	59
	NLOS	75
Urban Macro	LOS	93
	NLOS	365
Rural Macro	LOS	32
	NLOS	37

$$PDP(t) = \left\langle |h(t)|^2 \right\rangle_N, h(t) = \text{IFT}[S_{21}]$$

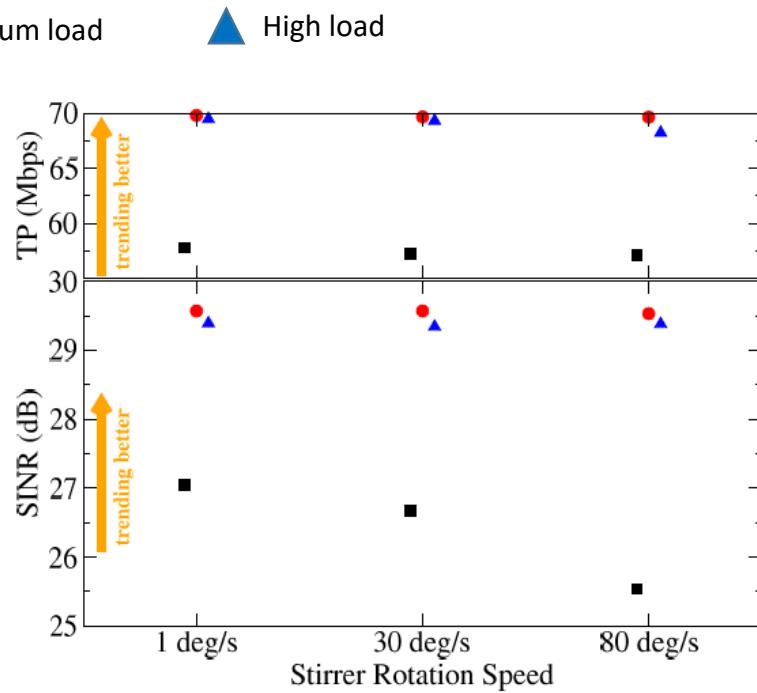
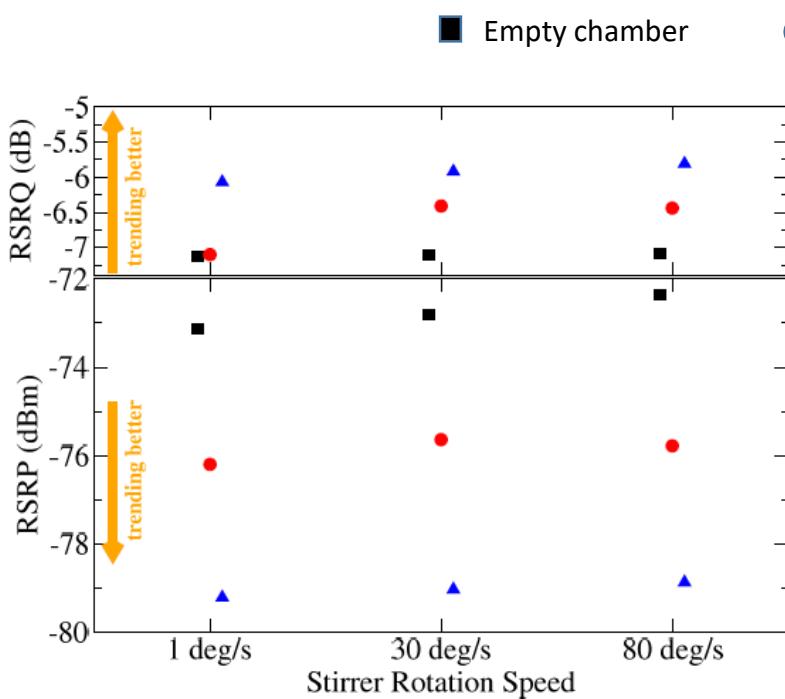
$$\tau_{\text{RMS}} = \sqrt{\frac{\int_0^\infty (t - \tau_{\text{ave}})^2 PDP(t) dt}{\int_0^\infty PDP(t) dt}}$$

$$\tau_{\text{ave}} = \frac{\int_0^\infty t PDP(t) dt}{\int_0^\infty PDP(t) dt}$$

D. Micheli, M. Barazzetta, F. Moglie and V. Mariani Primiani, "Power Boosting and Compensation During OTA Testing of a Real 4G LTE Base Station in Reverberation Chamber," in *IEEE Transactions on Electromagnetic Compatibility*, vol. 57, no. 4, pp. 623-634, Aug. 2015.



Example of a transmission quality test



- Reference signal received power (RSRP): is the average of the power of resource elements that carry cell-specific reference signals
- Reference signal received quality (RSRQ): is based on the ratio of RSRP and RSSI (total wideband received power)

- PDSCH net throughput: is the throughput measured at physical layer at the client in the data downlink channel, removing the re-transmissions of negatively acknowledged TBs.
- Signal to interference and noise ratio (SINR): is the ratio between the wanted part of the signal and the sum of interference and noise

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- D. Micheli, M. Barazzetta, F. Moglie and V. Mariani Primiani, "Power Boosting and Compensation During OTA Testing of a Real 4G LTE Base Station in Reverberation Chamber," in *IEEE Transactions on Electromagnetic Compatibility*, vol. 57, no. 4, pp. 623-634, Aug. 2015.
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Many, many others .. sorry for missing.

Dedicated sessions at Conferences

- EMC Europe 2018, INTERNATIONAL SYMPOSIUM AND EXHIBITION ON ELECTROMAGNETIC COMPATIBILITY, 27-30 AUG., AMSTERDAM.
- IEEE EMCS Symposium on ELECTROMAGNETIC COMPATIBILITY, SIGNAL & POWER INTEGRITY, July 30 - August 3, 2018, Long Beach (CA).

